

# Design of Combinational Fractal Microstrip Patch Antenna using Two Feeding Techniques

Dr. Yogesh Bhomia<sup>1</sup>, Arushi Bhardwaj<sup>2</sup>, Ruhika Badhan<sup>3</sup>

**Abstract** – The great advances in communication system led to the demand of a multi-band, larger gain, compact, low-profile fractal antennas to hold up multiple wireless applications. The self-similarity property of Fractal antenna is advantageous in generating multiple frequencies or enhancing bandwidth. This paper describes the design and simulation of combination of sierpinski and crown shaped fractal antenna up to third iteration on IE3D software The propounded antenna is designed on 1.6mm thick FR4 substrate with dielectric constant,  $\epsilon_r$  of 4.4 and is fed with 50 ohms for two types of feeding and is mounted above the ground plane at a height of 6 mm.

Details of the measured and simulated results of the individual iterations is presented and discussed.

**Keywords** – Feed, Fractal, Microstrip Antenna, Sierpinski Fasket.

## I. INTRODUCTION

In communication system, variety of microstrip antennas are being utilized, the most usual of which is microstrip patch antenna [12]. A patch antenna is a narrow band, wide-beam, low-profile, light-weight, conformal-shaped antenna fabricated by etching the antenna element pattern in metal trace joined to an insulating dielectric substrate. It is incorporated with a flat rectangular sheet or “patch” of metal, mounted over a larger sheet of metal called a ground plane. A patch antenna is mainly constructed on a dielectric substrate employing the same materials & lithography techniques in order to make printed circuit boards. Microstrip or patch antennas [6] are becoming more and more useful because they can be printed directly onto a circuit board. Furthermore, they are becoming ubiquitous within the mobile phone market [1]. These are somewhat inexpensive to manufacture and design because of the simple 2-dimensional physical geometry. These are also proficient of dual & triple frequency operations. These are highly efficient, easily integrated to circuits, compatible to the planer & non-planer surfaces and MMIC design. All these features make microstrip antennas widely implemented in many applications, such as high performance aircrafts, wireless communication satellite and missile applications. Fractal antennas [11] can be put to use in a variety of applications, especially where space is minimal. An exemplar illustrating the advantages of fractal in antenna system is the phased arrays, where fractals can diminish mutual coupling. Additionally, microstrip patch antennas are also subjected to some drawbacks, Narrow bandwidth being a serious curb. Different techniques [2] are proposed to improve it, and one of the methods proposed by various researchers is by cutting slots on it.

In this paper we have presented a design of microstrip Patch antenna using Crown & Sierpienksi fractal slots

[13], with an aim to achieve a smaller size antenna [4]. Target of this work is to design a microstrip patch antenna and carrying out results using commercial simulation software like IE3D. IE3D, from zeland software, Inc.[7], is an electromagnetic simulation and optimization software useful for circuit and antenna design. IE3D has a menu driven graphic interface for model generation with automatic meshing, and uses a field solver based on full wave , method-of-moments to solve current distribution on 3D and multilayer structures of general shape. IE3D usually focuses on general planar and 3D metallic structures in layered dielectric environments.

## II. DESIGN OF FRACTAL ANTENNA

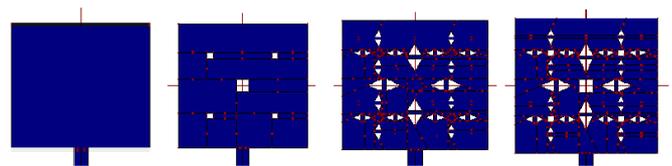


Fig.1. Sierpinski & crown combinational with Reference, 1<sup>st</sup> iteration, 2<sup>nd</sup> iteration, 3<sup>rd</sup> iteration

### Design Parameters

The transmission line model is used to design rectangular microstrip fractal antenna.

### Patch Width and Length

The first step is to design the patch is choosing a suitable di-electric substrate of suitable thickness. For rectangular microstrip antenna, the width  $W$  and the length  $L$  depends on the resonant frequency  $f_r$  and the parameters of the substrate employed [9]

To design the rectangular patch width of the antenna is given by-

### Width of the Patch

$$W = \frac{c}{f_r \sqrt{\epsilon_r + 1}} \quad (1)$$

Where,  $c$  is the speed of light,

$f_r$  is the resonant frequency.

### Effective Dielectric Constant

$$\epsilon_{\text{reff}} = (\epsilon_r + 1)/2 + (\epsilon_r - 1)/2 \sqrt{1 + 12 h/w} \quad (2)$$

Where,  $\epsilon_{\text{reff}}$  is the effective dielectric constant,

$\epsilon_r$  is the dielectric constant,

$h$  is the height of the substrate,

$W$  is the width of the patch.

### Taking into Account the Fringing Effect

The fringing fields along the width of the structure are taken as radiating slots and the patch antenna is electrically seen to be a bit larger than its physical size.

$$\Delta L = 0.412h \frac{(\epsilon_{\text{reff}} + 0.3) \left( \frac{W}{h} + 0.264 \right)}{(\epsilon_{\text{reff}} + 0.258) \left( \frac{W}{h} + 0.8 \right)} \quad (3)$$



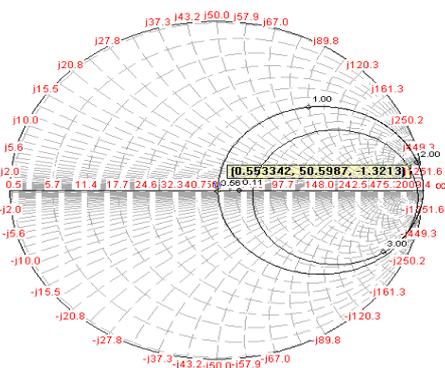


Fig. 7. Input impedance loci using smith chart of second iteration

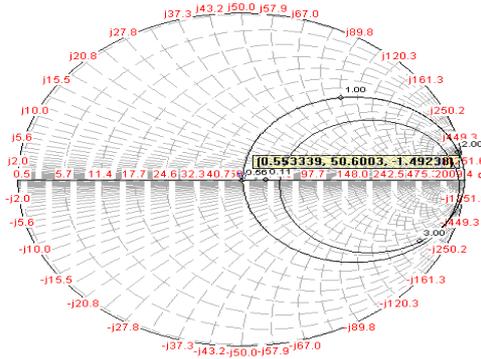


Fig. 11. Input impedance loci using smith chart of third iteration

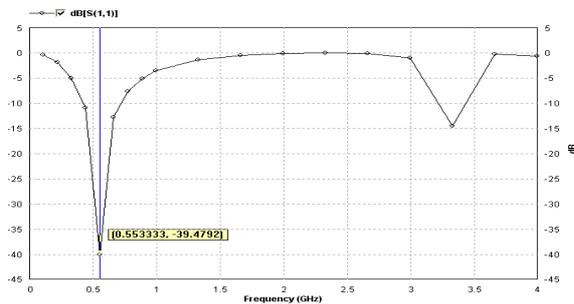


Fig. 8. S11 parameter for 2<sup>nd</sup> iteration antenna

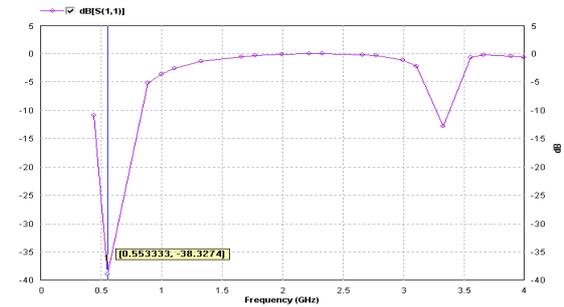


Fig.12. S11 parameter

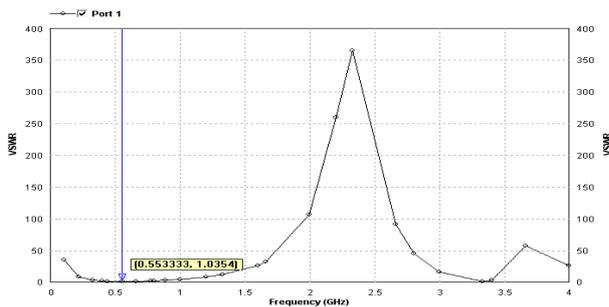


Fig. 9. VSWR

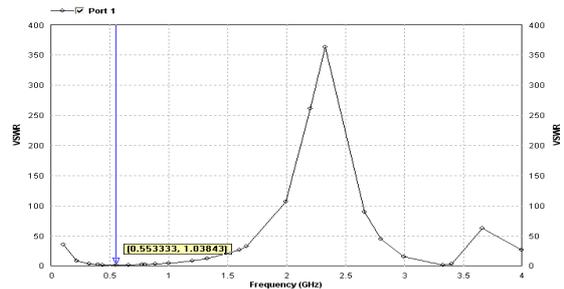


Fig. 13. VSWR

The figure 6 shows the model of 2<sup>nd</sup> iteration antenna. Figure 7, 8, 9 shows the results from IE3D simulations, as may be noted that the antenna has a good return loss of magnitude -39.4792 at the frequency 0.553333.

### Third Iteration

The geometry of 3<sup>rd</sup> iteration of proposed microstrip patch antenna using Crown & Sierpienksi fractal slots presented. Third iteration is obtained by combining the geometry of both iterations 1 & 2.

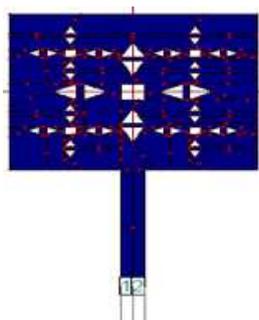


Fig. 10. The model for 3<sup>rd</sup> iteration

The figure 10 shows the model of 3<sup>rd</sup> iteration antenna. Figure 11, 12, 13 shows the results from IE3D simulations, as may be noted that the antenna has a good return loss of magnitude -38.3214 at the frequency 0.553333.

### III. PROBE FEED

#### Iteration 1

The figure 14 shows the model of Iteration 1 for probe feed antenna & the input impedance loci using smith chart. The points for the probe feed are (x, y) = 3, 9).

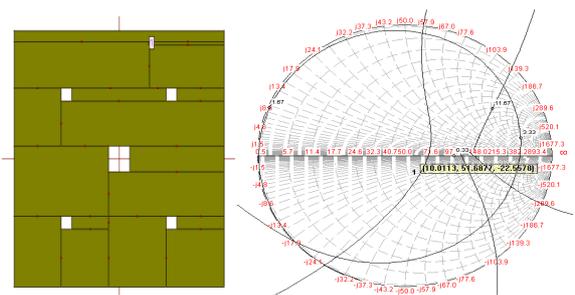


Fig. 14. Model for 1<sup>st</sup> iteration, I/P impedance loci using smith chart

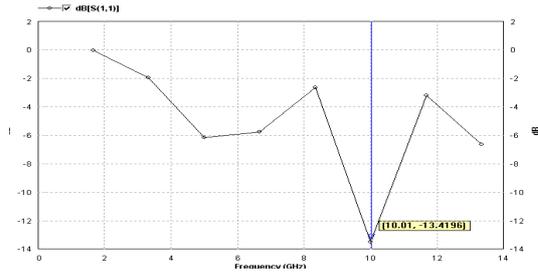


Fig. 15. S11 Parameter

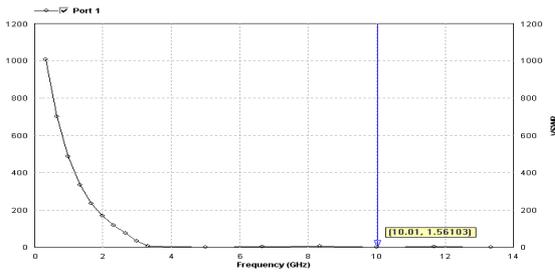


Fig.16. VSWR

**Iteration 2**

Fig. 17 shows the location of the probe feed. The probe feed location is adjusted so as to connect to the metallic portion of the patch & where a max. return loss is observed. Fig. 18, 19 presents the results from IE3D simulations.

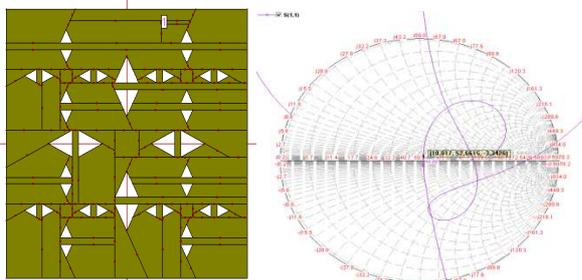


Fig. 17. Model for 2<sup>nd</sup> iteration, smith chart

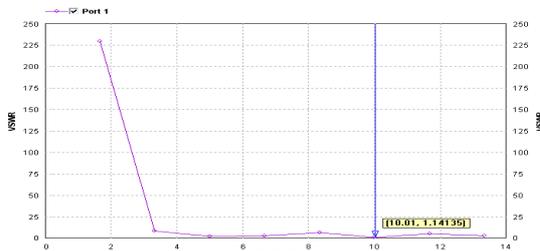


Fig. 18. VSWR

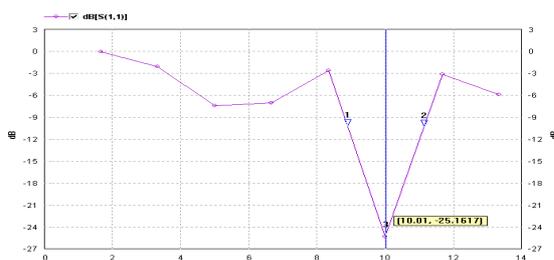


Fig.19. S11 Parameter

**Iteration 3**

Fig. 20 shows Iteration 3 for probe feed & the location of feed. The probe feed location is adjusted so as to connect to the metallic portion of the patch and where a maximum return loss is observed. Fig. 21 and 22 shows the results from IE3D simulations.

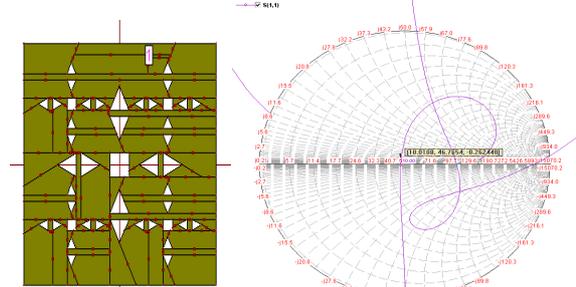


Fig. 20. Model for 3<sup>rd</sup> iteration, smith chart

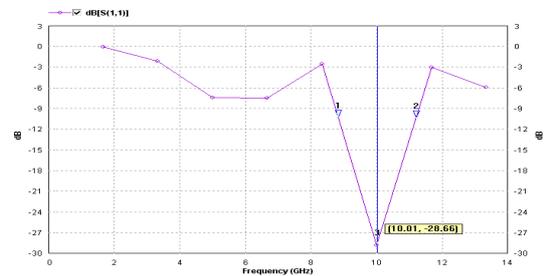


Fig. 21. S11 Parameter

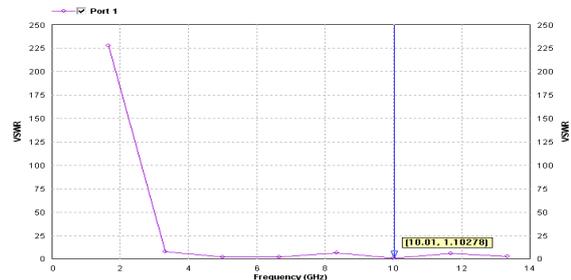


Fig. 22. VSWR

Table 1. Comparison of different results of Iteration 1, 2&3

TRANSMISSION –LINE FEED			
Types	Iteration 1	Iteration 2	Iteration 3
Resonant frequency	0.553333	0.553333	0.553333
Bandwidth	53.01%	53.02%	53.04%
VSWR	1.03265	1.0354	1.03843
Return Loss	-40.6734	-39.4792	-38.3214

PROBE FEED			
Types	Iteration 1	Iteration 2	Iteration 3
Resonant frequency	10.01	10.01	10.01
Bandwidth	10.7%	22.37%	23.78%
VSWR	1.56103	1.14135	1.10278
Return Loss	-13.4196	-25.1617	-28.66

**IV. CONCLUSION**

Design is simulated and result of the designed antenna is obtained using two different feeding techniques [14]. Traditionally, the wideband antennas (spiral and log-periodic) and arrays [10] designed using Euclidean

geometry [8]. The results demonstrated a maximum patch size reduction by the propounded fractal antennas, without deteriorating antenna's performance, such as the return loss and radiation pattern, vswr. This size reduction technique is loading the inductive elements along the patch edges, and loading self-similar slots inside the patch, to increase the length of the current path. The basis of the maintenance of the antenna radiation patterns is the self-similarities and centro symmetry of the fractal shapes [9].

The main advantages of the propounded method are: (i) size reduction (ii) maintained radiation patterns, (iii) wider operating frequency bandwidth (iv) simple and easy design. Thence, this is the most effective technique propounded for the size reduction of microstrip patch antennas so far. The small-sized patches obtained from this technique can be used in integrated low-profile wireless communication systems positively. In future fractal microstrip antenna reduced patch size and improved bandwidth can be achieved definitely.

The simulated results indicate that the antenna is suitable for wireless LAN [14] & networkable PDA, satellite communication (Direct broadcast TV), vehicle speed detection.

## V. RESULTS AND DISCUSSION

The propounded antenna has been simulated by using IE3D by Zealand Software Inc.[7] . It is considered as a standard for electromagnetic simulation packages. The prime formulation of the IE3D is an integral equation that is obtained through the use of Green's functions. This paper presented a square microstrip patch antenna combining [13] Crown and Sierpienksi fractal slots is fabricated on a FR4 substrate of relative permittivity of 4.4 & thickness 1.6 mm. It is mounted above the ground plane at height of 6 mm. [5]

Table 1 shows the variation of return loss with frequency, VSWR and Bandwidth for iteration I, II and III for transmission line feed & coaxial feed. [3]

Plot result shows resonant frequency 0.553333 GHz for transmission line feed & 10.01 GHz for coaxial feed. Return loss for the iteration I, II & III is -40.67, -39.47 & -38.32 respectively for transmission line feed. Return loss for the iteration I, II & III is -13.41, -25.16 & -28.66 respectively for coaxial feed.

## VI. REFERENCES

- [1] James, J.R. and Hall, P.S.: 'Handbook of Microstrip Antennas'(Peter Peregrinus)
- [2] Constantine A. Balanis: 'Antenna Theory, Analysis and Design' (John Wiley & Sons)
- [3] Bhomia Y., Chaturvedi A., Yadav D., "Truncated Tip Triangular Microstrip Patch Antenna", Proc. IEEE. Int Symp. Antenna Propag., vol. 2, pp. 212 -214, 2010.
- [4] Kai-Fong Lee, Kwai-Man Luk, Jashwant S. Dahele, 'Characteristics of the Equilateral Triangular Patch Antenna' IEEE Tran. Antennas Propag.1988; 36; 1510.
- [5] C. L. Mak, K. M. Luk, K. F. Lee, ' Wideband Triangular Patch Antenna' IEE. Proc. Microwave Antennas Propag. 1999.
- [6] I. J. Bahl and P. Bhartia, 'Microstrip Antennas, Artech House, Dedham, MA, 1980.
- [7] IE3D [1] by Zeland software Inc.

- [8] H. O. Peitgen, H. Jurgens, and D. Saupe, Chaos and Fractals. New York: Springer-Verlag, 1990.
- [9] M. F. Barnsley, R. L. Devaney, B. B. Mandelbrot, H. O. Peitgen, D.Saupe, R. F. Voss, Y. Fisher, and M. Mc Guire, The Science of Fractal Images. New York: Springer-Verlag, 1988.
- [10] Yogesh Bhomia et.al., "V-Slotted Triangular Microstrip Patch Antenna", Int. Journal of Electronics Engineering, vol. 2, no.1, pp. 21-23, 2010
- [11] K. J. Vinoy, J. K. Abraham, and V. K. Varadan, "On the Relationship Between Fractal Dimension and the Performance of Multi-Resonant Dipole Antennas using Koch Curves, " IEEE Transactions on Antennas and Propagation, AP-5 1, 9, 2003, pp.2296-2303.
- [12] Bhomia, Y., A. Kajla, and D. Yadav, "Slotted right angle triangular microstrip patch antenna, " International Journal of Electronic
- [13] Bhomia Y., Chaturvedi A., Yadav D., " Microstrip Patch Antenna Combining Crown and Sierpinski Fractal-Shapes ", Proc. IEEE. Int Symp. Antenna Propag., vol. 2, pp. 212 -214, 2010
- [14] Ankur Aggarwal , M.V.Kartikeyan, "Design of Sierpinski Carpet Antenna using two different feeding mechanism for WLAN applications", IEEE, 2010.
- [15] J. T. Bernhard, Reconfigurable Antennas. San Rafael, CA, USA: Morgan & Claypool, 2007 Engineering Research, Vol. 2, No. 3, 393-398, 2010.

## AUTHORS' PROFILES



**Dr. Yogesh Bhomia** received the B.E. degree from IETE, New Delhi in Electronics & telecommunication and M.E. in digital communication from M.B.M. College, Jodhpur and also received Ph.D. degree from Bhagwant University, Ajmer, Rajasthan, India. He is currently working as a Director Principal in SSCET, Badhani. He has been published more than 25 papers in International Journals. He is reviewer of various Journals and also member of various IEEE technical Committees. He has written a book on Electro Magnetic Field Theory.



**Er. Arushi Bhardwaj** received the B.Tech degree from PTU, Punjab in Electronics & communication and M.Tech in ECE from PTU, Punjab. She is currently working as an Assistant Professor in SSCET, Badhani in Electronics & Communication Engineering



**Er. Ruhika Badhan** received the B.E. degree from PTU, Punjab in Electronics & telecommunication and M.Tech in ECE from PTU, Punjab.